

# Mobile Communications

TCS 455

**Dr. Prapun Suksompong**

[prapun@siit.tu.ac.th](mailto:prapun@siit.tu.ac.th)

**Lecture 25**

**Office Hours:**

**BKD 3601-7**

**Tuesday 14:00-16:00**

**Thursday 9:30-11:30**

# Announcements

- Read the following from the SIIT online lecture note system
  - Section 1.2 from [Bahai, *Multi-carrier Digital Communications: Theory And Applications Of OFDM*, 2002]
  - Theory of PN Codes from [Karim and Sarraf, *W-CDMA and cdma2000 for 3G Mobile Networks*, 2002]
- Check the course web site for links to interesting resources about DFT.
- Communication Group Meeting: Presentation starts at 2:30 PM.
  - Move to the 4<sup>th</sup> floor quickly after lecture.

# Chapter 5

## OFDM

**Office Hours:**  
**BKD 3601-7**  
**Tuesday 14:00-16:00**  
**Thursday 9:30-11:30**

# Three steps towards modern OFDM

1. Mitigate Multipath (ISI) → Multicarrier modulation (FDM)
  2. Gain Spectral Efficiency → Orthogonality of the carriers
  3. Achieve Efficient Implementation → FFT and IFFT
- Completely eliminate ISI and ICI → Cyclic prefix

# Chapter 5

## OFDM

Cyclic Prefix (CP)

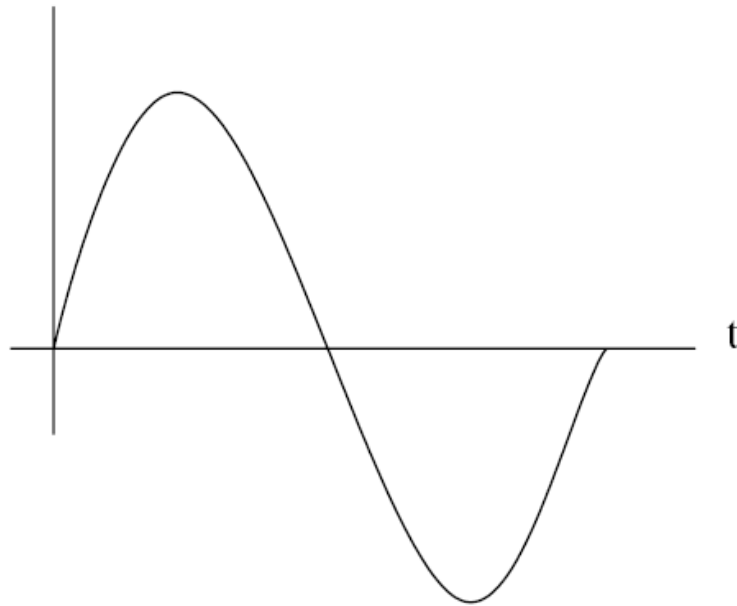
**Office Hours:**

**BKD 3601-7**

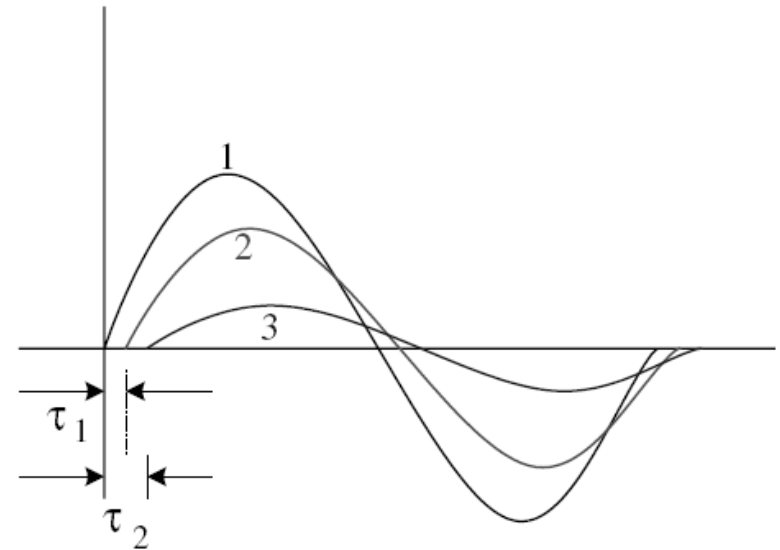
**Tuesday 14:00-16:00**

**Thursday 9:30-11:30**

# Cyclic Prefix: Motivation (1)



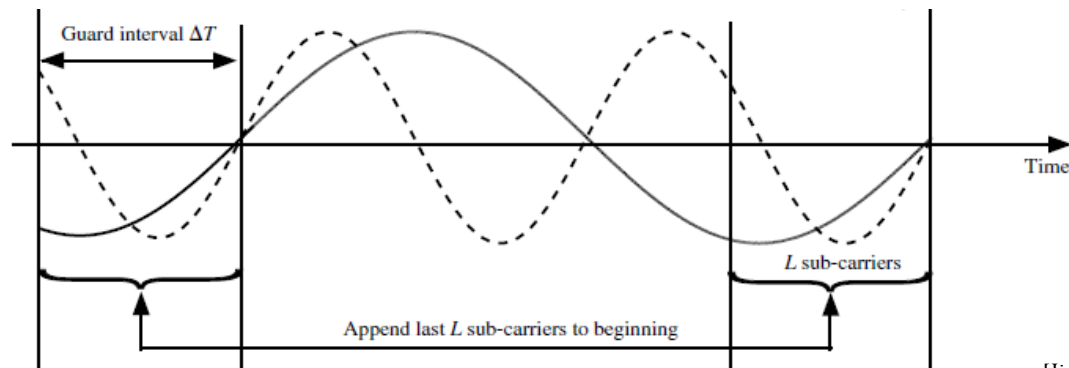
Transmitted  
Signal



Received Signal

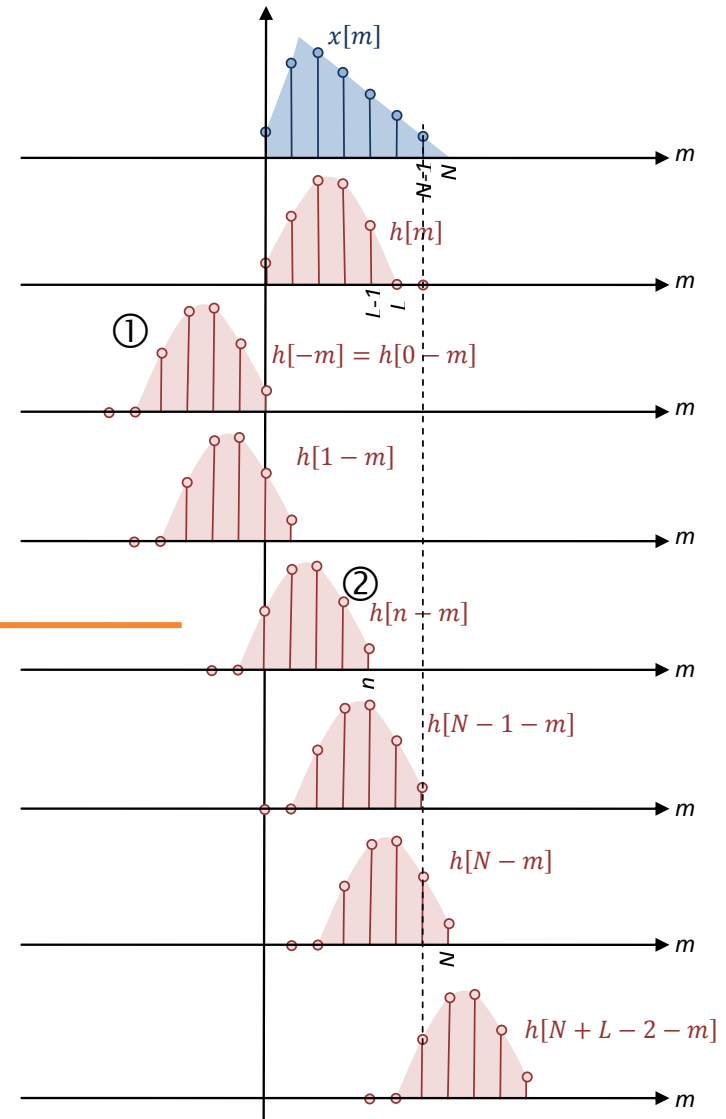
# Cyclic Prefix: Motivation (2)

- When the number of sub-carriers increases, the OFDM symbol duration  $T_s$  becomes large compared to the duration of the impulse response  $\tau_{\max}$  of the channel, and the amount of ISI reduces.
- Can we “eliminate” the multipath (**ISI**) problem?
- To reduce the ISI, add **guard interval** larger than that of the estimated delay spread.
- If the guard interval is left empty, the orthogonality of the sub-carriers no longer holds, i.e., **ICI** (inter-channel interference) still exists.
- To prevent both the ISI as well as the ICI, OFDM symbol is **cyclically extended** into the guard interval.



# Convolution

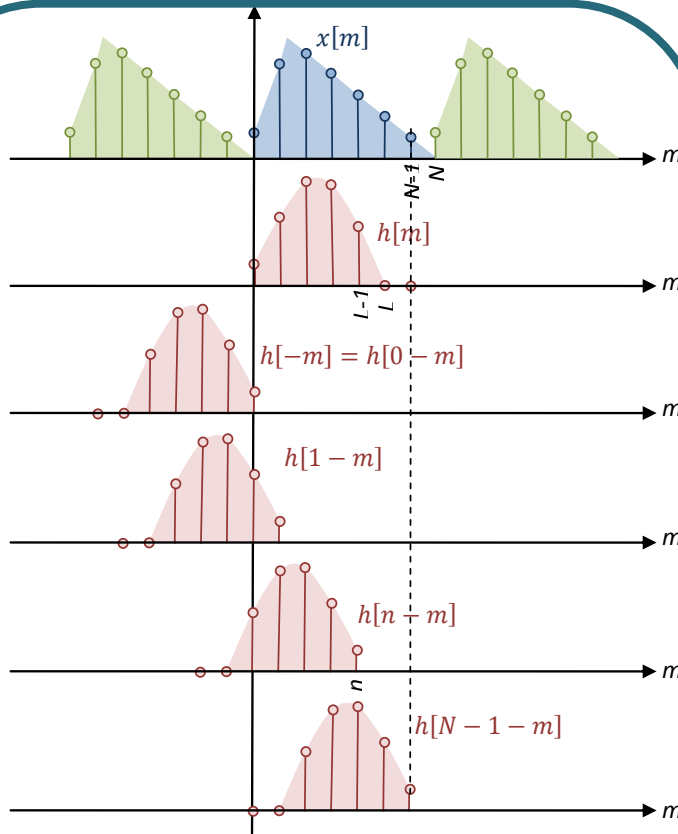
- ① Flip
- ② Shift
- ③ Multiply
- ④ Add



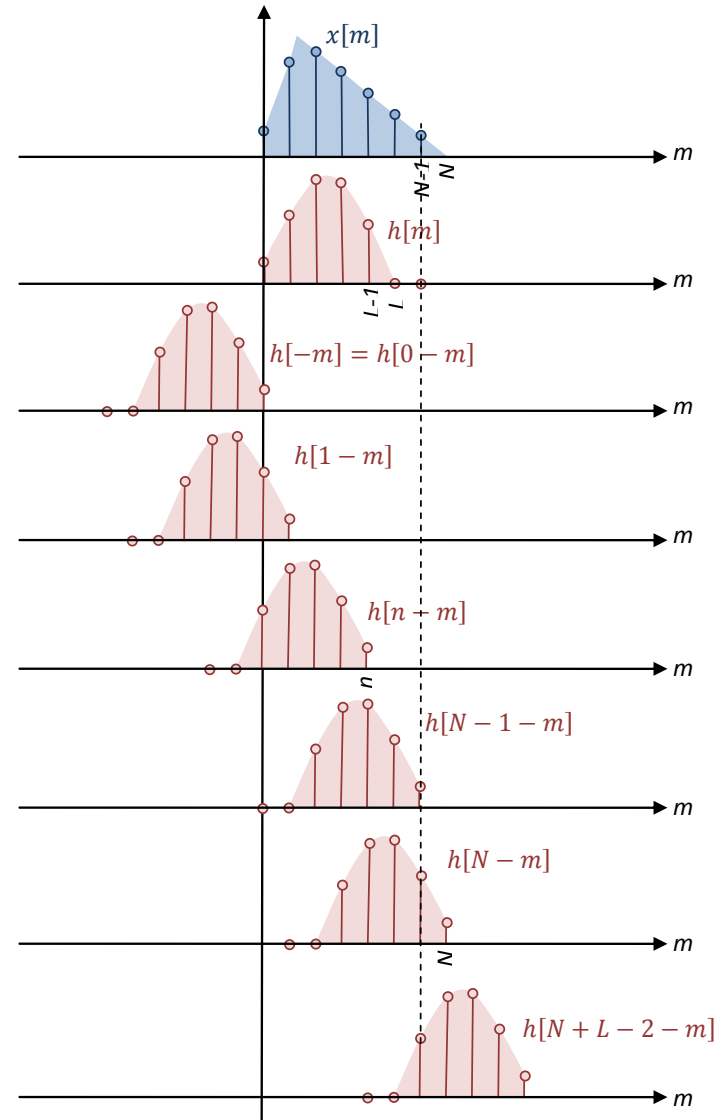
$$\{x * h\}[n] = \sum_m x[m]h[n-m]$$



# Circular Convolution



Replicate  $x$  (now it looks periodic)  
Then, perform the usual convolution  
only on  $n = 0$  to  $N-1$



# Circular Convolution: Example

Find

$$[1 \ 2 \ 3] * [4 \ 5 \ 6]$$

$$[1 \ 2 \ 3] \otimes [4 \ 5 \ 6]$$

$$[1 \ 2 \ 3 \ 0 \ 0] \otimes [4 \ 5 \ 6 \ 0 \ 0]$$

# Discussion

- Circular convolution can be used to find the regular convolution by zero-padding.
- In modern OFDM, it is another way around.
- CTFT: convolution in time domain corresponds to multiplication in frequency domain.
- DFT: circular convolution in (discrete) time domain corresponds to multiplication in (discrete) frequency domain.
  - We want to have multiplication in frequency domain.
  - So, we want circular convolution and not the regular convolution.
- Real channel does regular convolution.
- With cyclic prefix, regular convolution can be used to create circular convolution.



# Circular Convolution: Key Properties

- Consider an  $N$ -point signal  $x[n]$
- **Cyclic Prefix (CP) insertion:** If  $x[n]$  is extended by copying the last  $v$  samples of the symbols at the beginning of the symbol:

$$\hat{x}[n] = \begin{cases} x[n], & 0 \leq n \leq N-1 \\ x[n+N], & -v \leq n \leq -1 \end{cases}$$

- Key Property 1:

$$\{h \circledast x\}[n] = (h * \hat{x})[n] \text{ for } 0 \leq n \leq N-1$$

- Key Property 2:

$$\{h \circledast x\}[n] \xrightarrow{\text{FFT}} H_k X_k$$

# OFDM with CP for Channel w/ Memory

- We want to send  $N$  samples  $S_0, S_1, \dots, S_{N-1}$  across noisy channel with memory.

- First apply IFFT:  $S_k \xrightarrow{\text{IFFT}} s[n]$

- Then, add cyclic prefix

$$\hat{s} = [s[N - \nu], \dots, s[N - 1], s[0], \dots, s[N - 1]]$$

- This is inputted to the channel.

- The output is

$$y[n] = [p[N - \nu], \dots, p[N - 1], r[0], \dots, r[N - 1]]$$

- Remove cyclic prefix to get  $r[n] = h[n] \otimes s[n] + w[n]$

- Then apply FFT:  $r[n] \xrightarrow{\text{FFT}} R_k$

- By circular convolution property of DFT,  $R_k = H_k S_k + W_k$

No ICI!

# Example: 802.11a

Parameter	IEEE 802.11a
Bandwidth	20 MHz
Number of sub-carriers $N_c$	52 (48 data + 4 pilots) (64 FFT)
Symbol duration $T_s$	4 $\mu$ s
Carrier spacing $F_s$	312.5 kHz = $\frac{1}{4-0.8[\mu\text{s}]}$
Guard time $T_g$	0.8 $\mu$ s
Modulation	BPSK, QPSK, 16-QAM, and 64-QAM
FEC coding	Convolutional with code rate 1/2 up to 3/4
Max. data rate	54 Mbit/s